



IEEE Custom Integrated Circuits Conference

14.8: A 16nm 2.2pJ/b High Efficiency Pulse-Position Modulated Resonance Transceiver for 15Mbps Multi-Source High-Speed Body-Area Sensor Network

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Outline

- Motivation
- The Proposed Sensor Node TRX Design
 - High Density Differential-Mode Pulse-Position-Modulation (DM-PPM)
 - Resonant Pulse Transmitter
 - Low Power Receiver
- Measurements & Conclusion

Outline

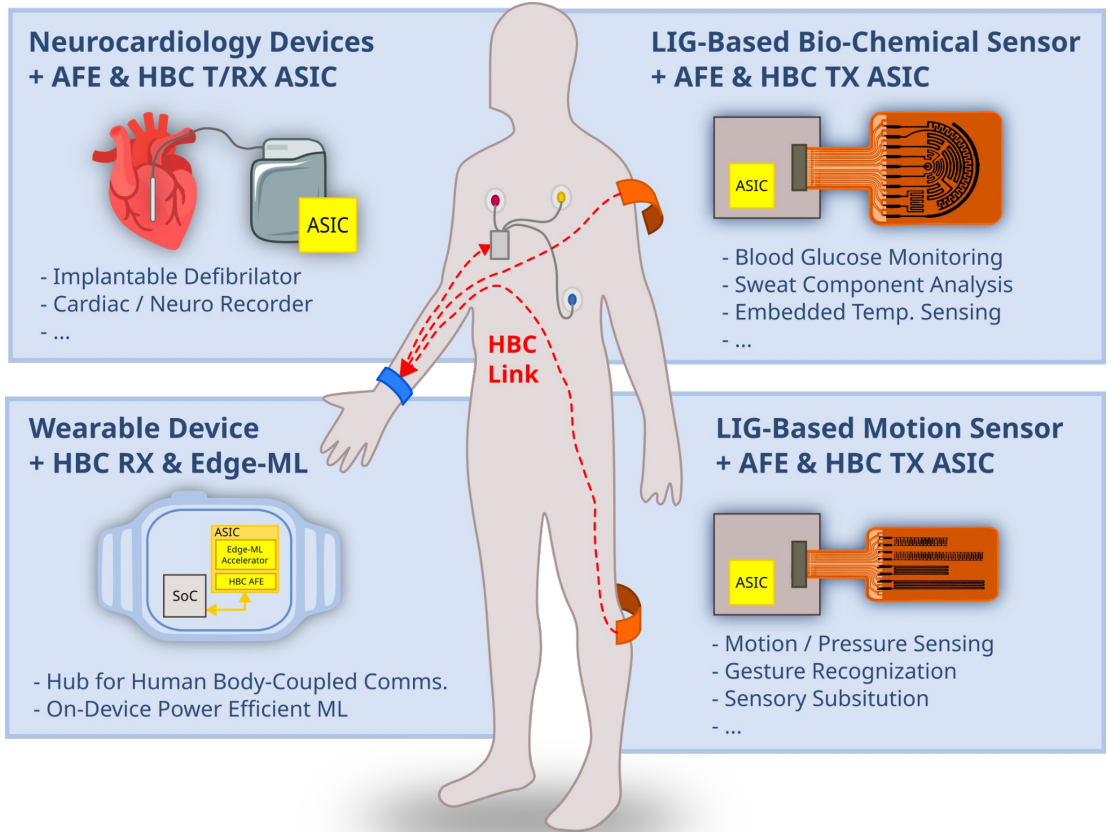
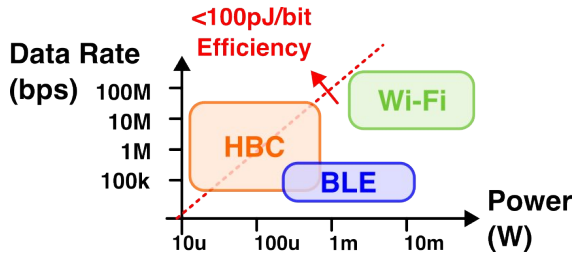
- **Motivation**
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Motivation

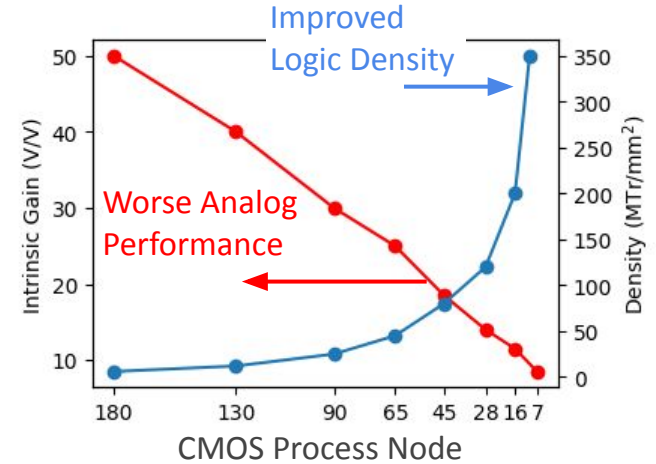
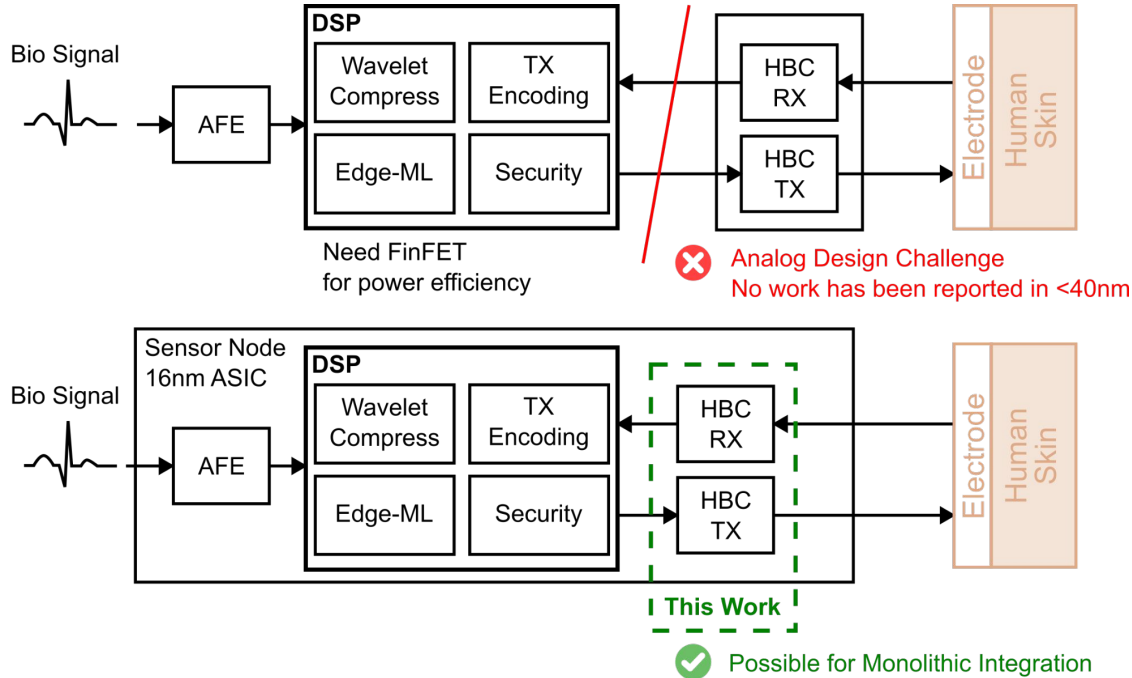
Communication of Sensor nodes

- Monitoring
- Bio Signature
- Neuro-Implant
- ...

HBC provides much higher energy efficiency than 2.4GHz radio.



Challenge: Integrating HBC AFE with Digital Processor



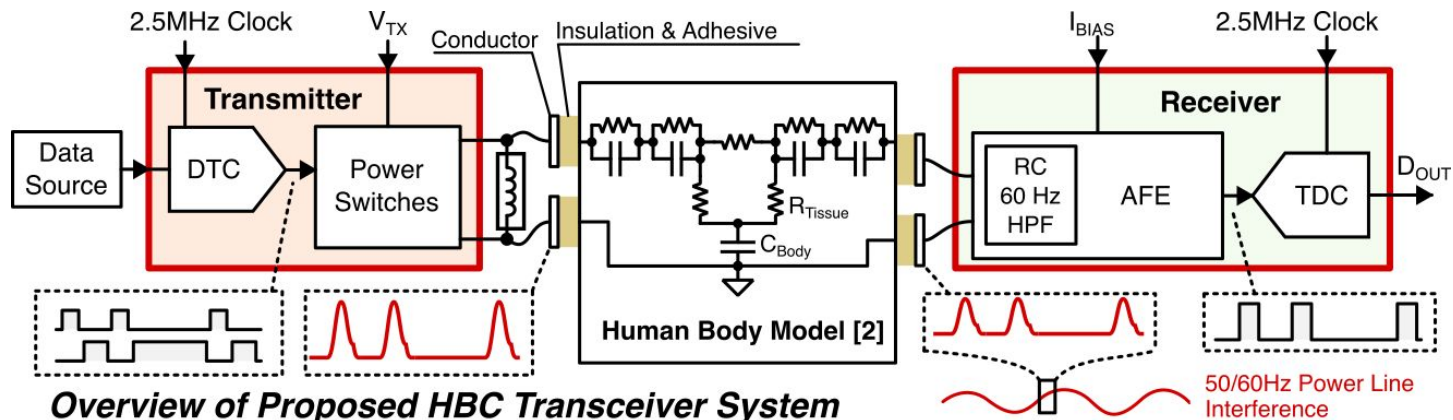
Challenge: Advanced nodes have high leakage and low headroom

No reported HBC transmitter in FinFET nodes

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Overview of the Proposed TRX System

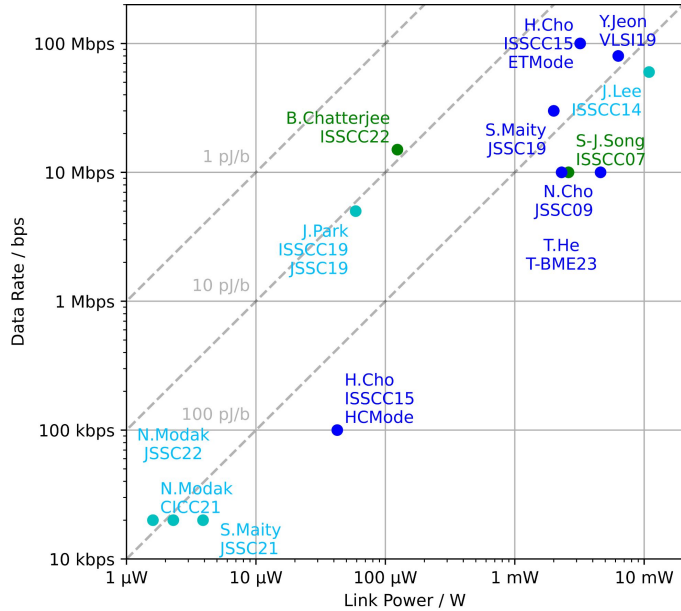


- Reduce pulse-per-bit → Better encoding scheme (DM-PPM)
- High efficiency TX → Use of High-Q Inductor
- High efficiency RX → FinFET-Optimized Circuit;
Power Management Techniques

Outline

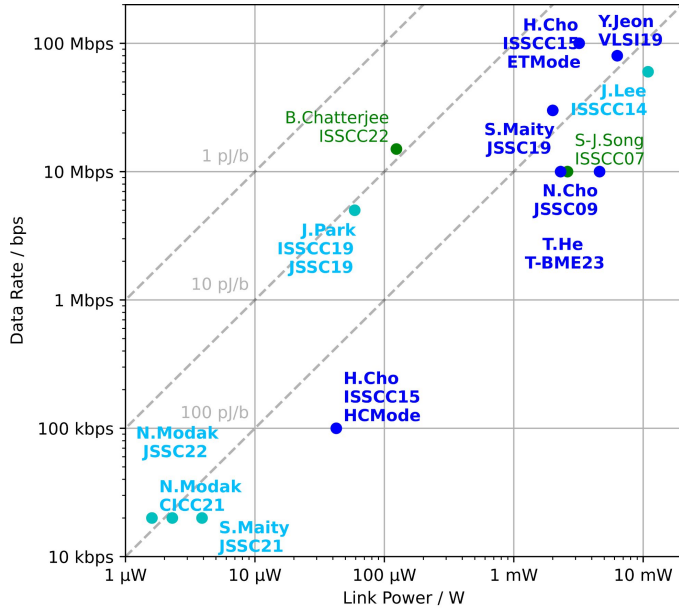
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Modulation Schemes in Prior Arts

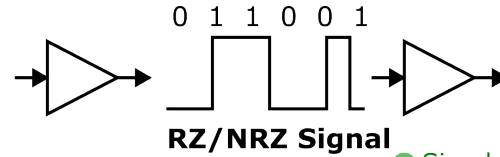


- Pulse Position Modulation (PPM)
- Simple Modulation (OOK / ASK)
- Advanced Modulation (PSK / FSK / PAM)

Modulation Schemes in Prior Arts

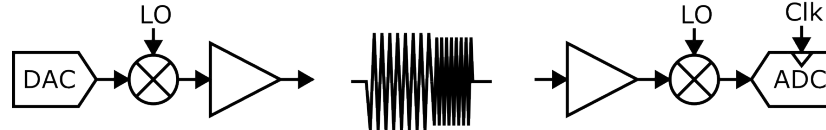


- Pulse Position Modulation (PPM)
- Simple Modulation (OOK / ASK)
- Advanced Modulation (PSK / FSK / PAM)



RZ/NRZ Signal

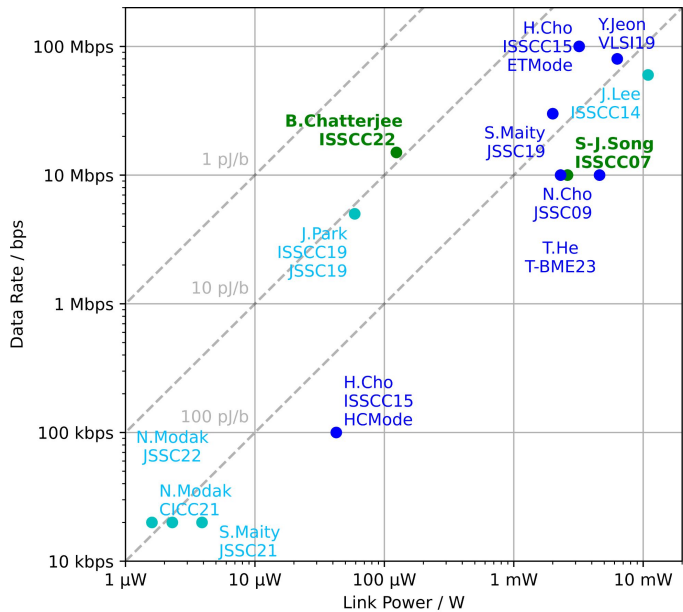
- ✓ Simple
- ✗ Low Data Rate
- ✗ Susceptible to Interference



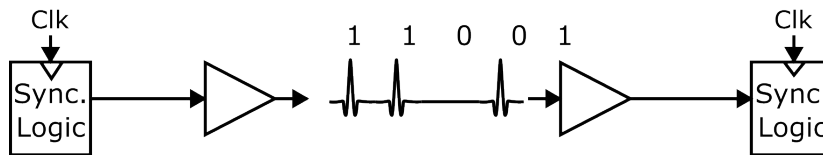
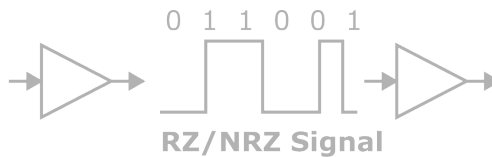
Modulated Signal (PSK/FSK/PAM)

- ✓ High Data Rate (>50Mbps)
- ✓ Good Efficiency for High Data Rate
- ✗ High Power for Mid/Low Data Rate
- ✗ High Design & Calibration Complexity

Modulation Schemes in Prior Arts



- Pulse Position Modulation (PPM)
- Simple Modulation (OOK / ASK)
- Advanced Modulation (PSK / FSK / PAM)

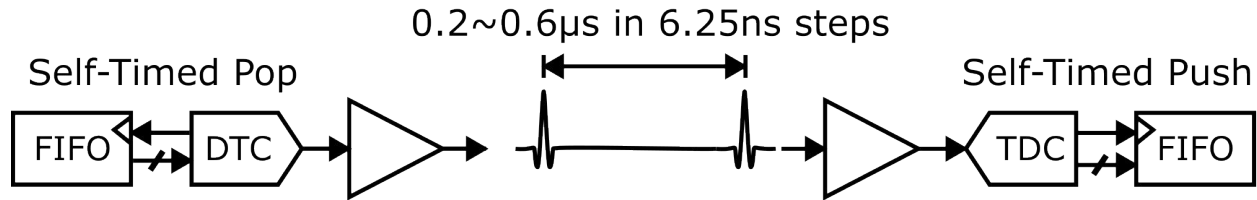


Pulse Position Modulation

- ✓ High Data Rate (~15Mbps)
- ✓ Low Power with Less Switching Activity
- ✓ Simple to Design & Calibrate
- ✗ Clock Rate limits Data Rate

High Density Differential-Mode Pulse-Position-Modulation (DM-PPM)

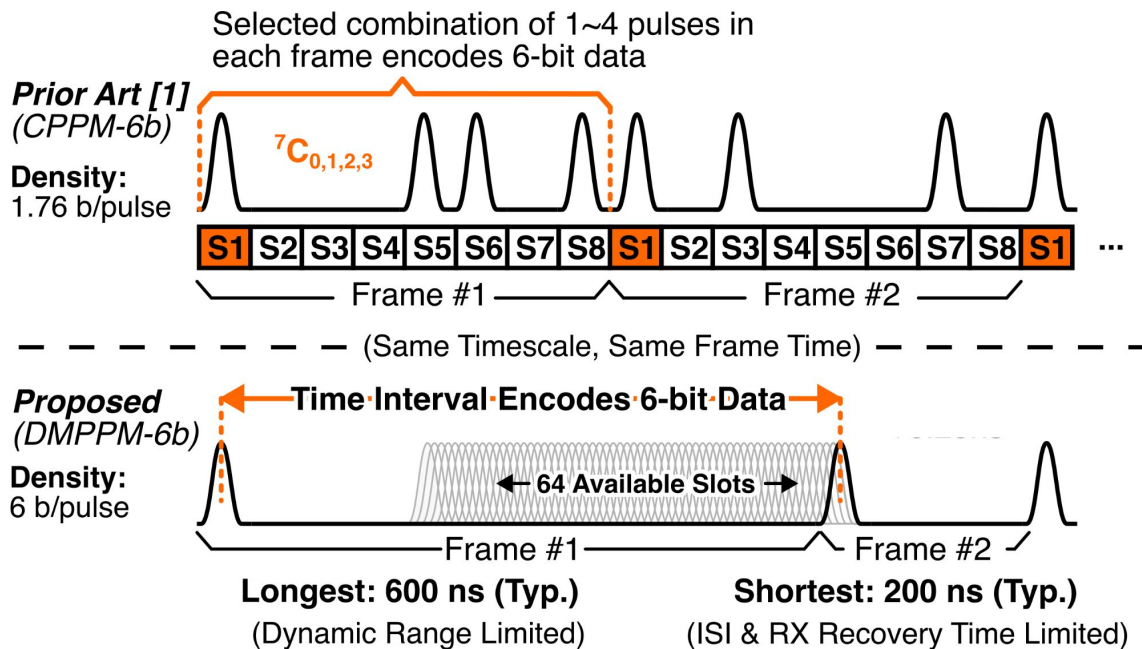
Proposed Fully Asynchronous Time-Domain Signaling



**Modulated Signal
(Pulse-Position Modulation)**

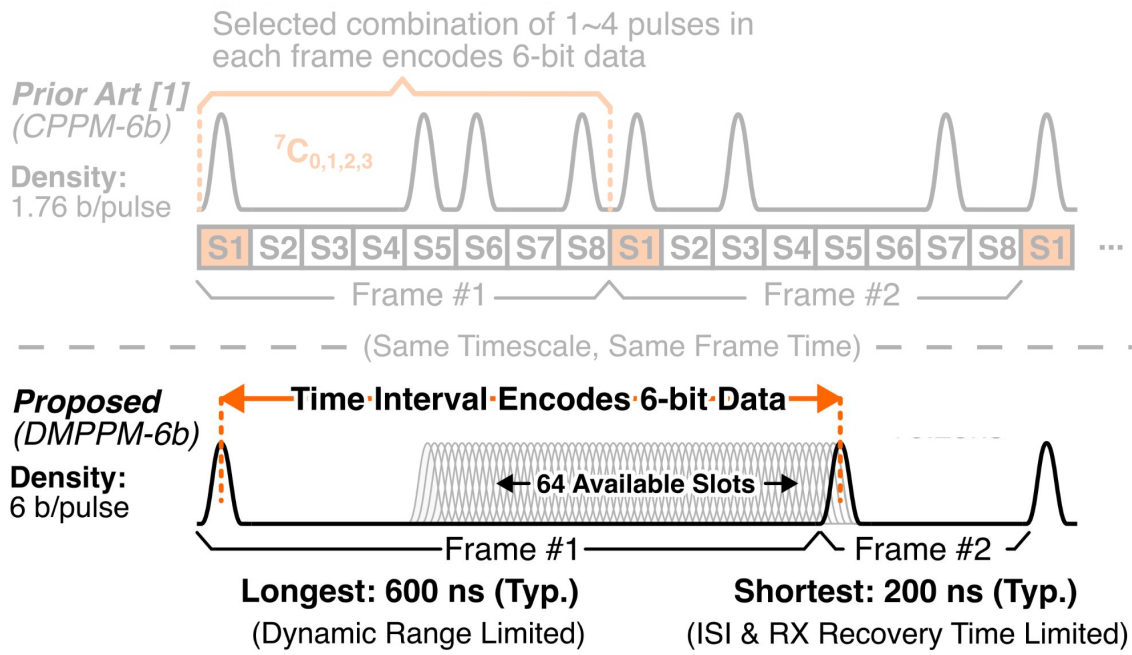
- ✓ High Data Rate (~15Mbps)
- ✓ Low Power with Even Fewer Switching Activity
- ✓ Simple to Design & Calibrate
- ✓ FinFET-Friendly Timing Generation
- ✓ Lower Clock Speed

High Density Differential-Mode Pulse-Position-Modulation (DM-PPM)



DTC+TDC enables sub-clock period resolution.

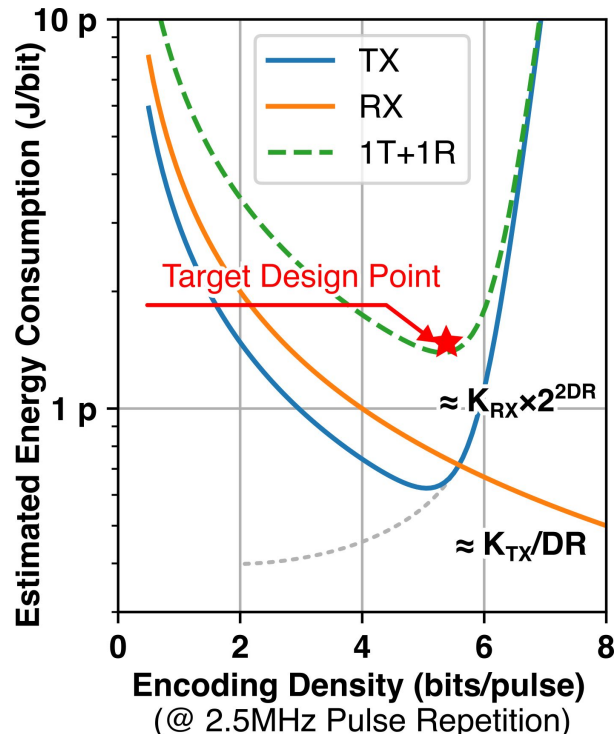
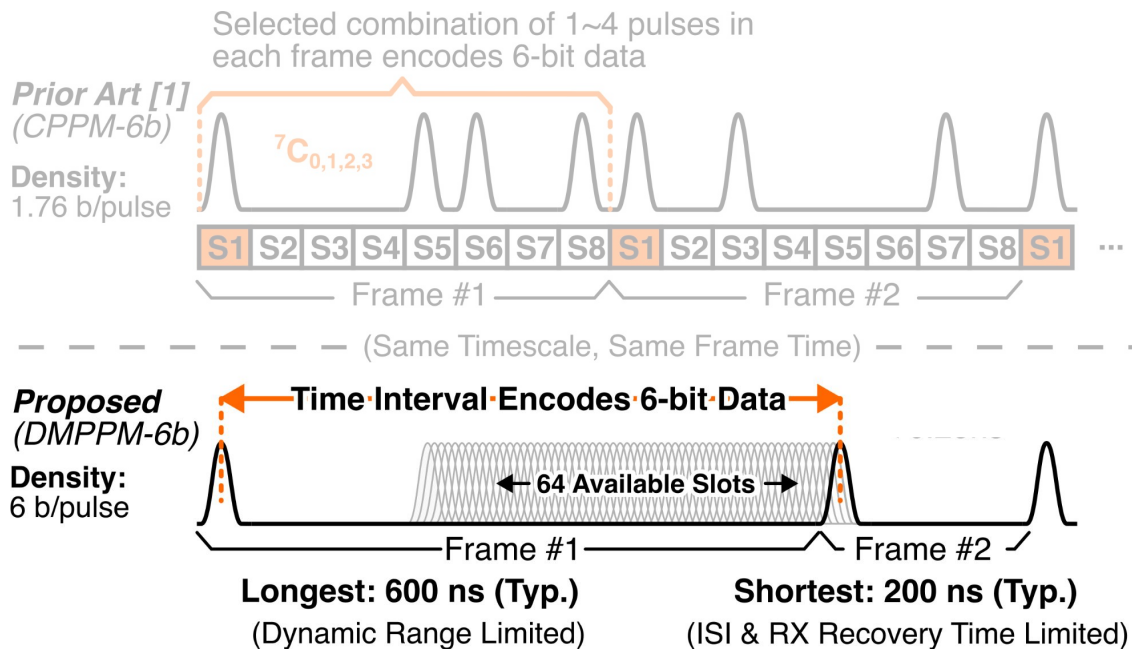
High Density Differential-Mode Pulse-Position-Modulation (DM-PPM)



DTC+TDC enables sub-clock period resolution.

How much resolution is appropriate?

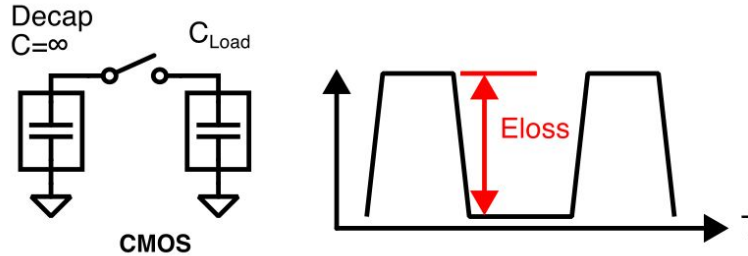
High Density Differential-Mode Pulse-Position-Modulation (DM-PPM)



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 - **Resonant Pulse Transmitter**
 - Low Power Receiver
- Measurements & Conclusion

Resonant Pulse Transmitter: Concept



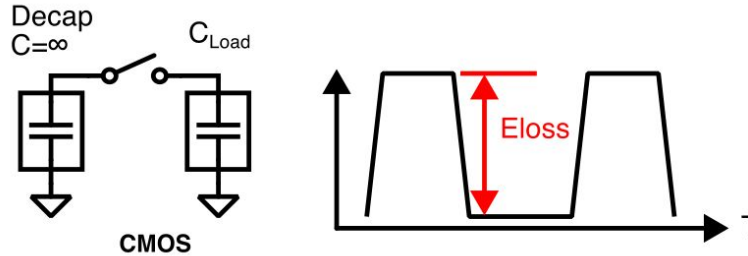
Energy Lost in 1 Cycle:

$$E_{loss} = E_{cloud}$$

Human Skin is a capacitive load
Driving with CMOS will cause 100% capacitive energy loss.

$$E_{loss} = CV^2 = \underline{\underline{25pJ}} @ 1.0V_{pp} \text{ per pulse}$$

Resonant Pulse Transmitter: Concept



Energy Lost in 1 Cycle:

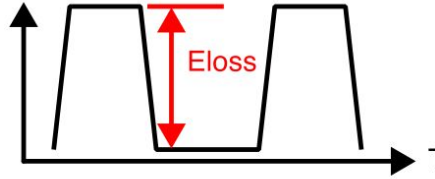
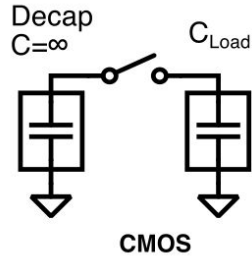
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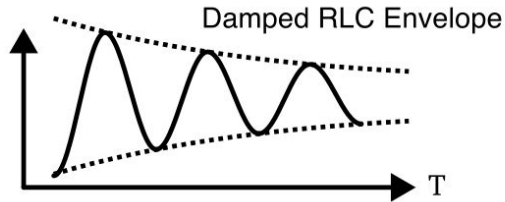
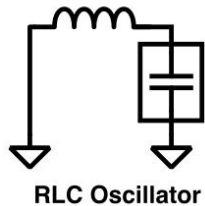
Can we do better?

Resonant Pulse Transmitter: Concept



Energy Lost in 1 Cycle:

$$E_{loss} = E_{cload}$$



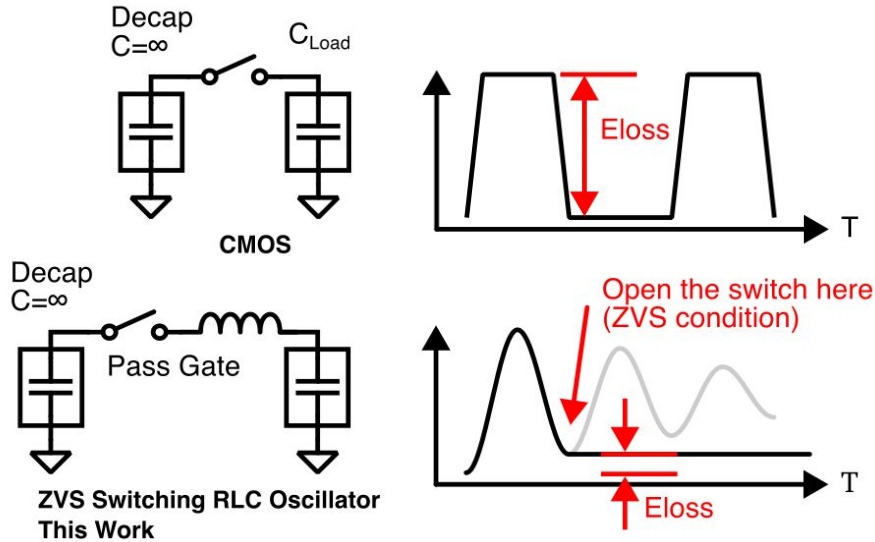
$$E_{loss} = E_{cload} (2\pi R / \omega_0 L)$$

$$= E_{cload} (2\pi / Q) \ll E_{cload}$$

Inductor can create oscillation on a capacitor with lower loss per cycle.

$$E_{loss} = E_{cap} \times (2\pi / Q) < 0.13 \times E_{cap} = \underline{\underline{3.2pJ}} @ 1.0V_{pp} \text{ per pulse, } Q=50$$

Resonant Pulse Transmitter: Concept



Energy Lost in 1 Cycle:

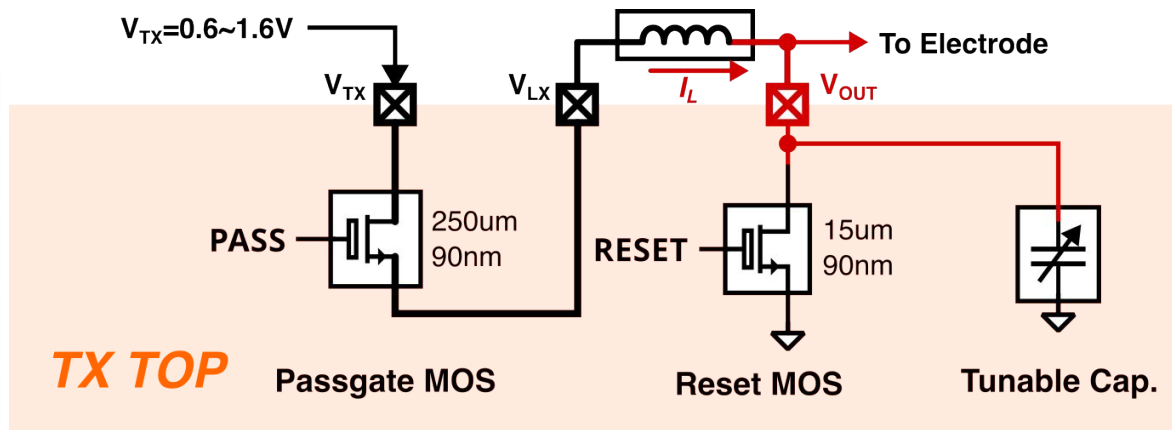
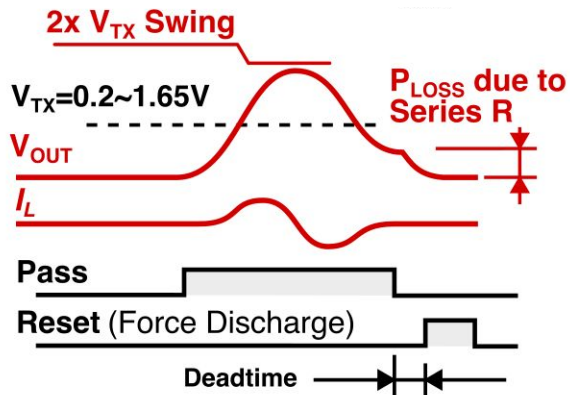
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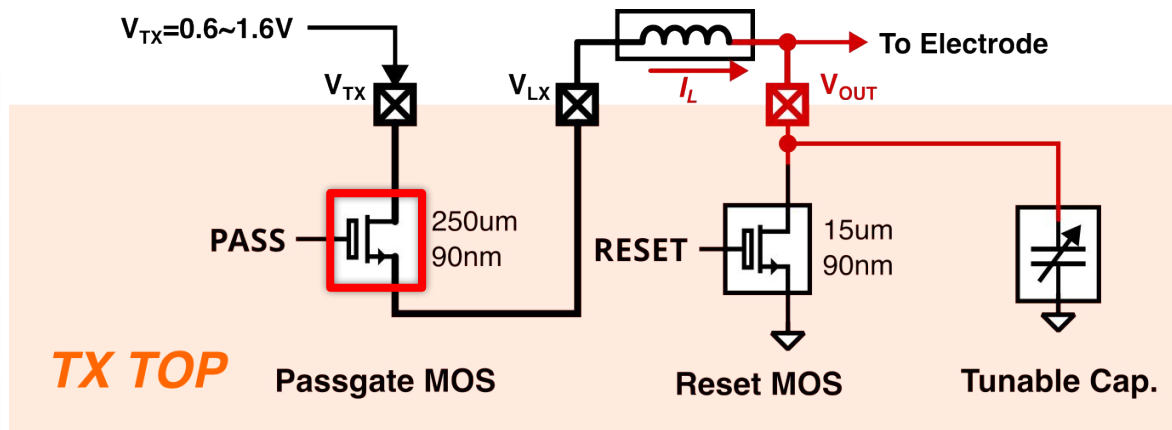
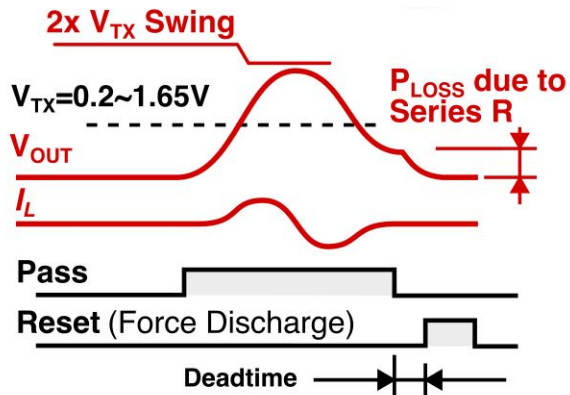
$$= E_{cload} (2\pi / Q) \ll E_{cload}$$

The switch can open/close at the peak/valley of the oscillation. without capacitive energy loss (ZVS condition)

Resonant Pulse Transmitter: Simplified Diagram

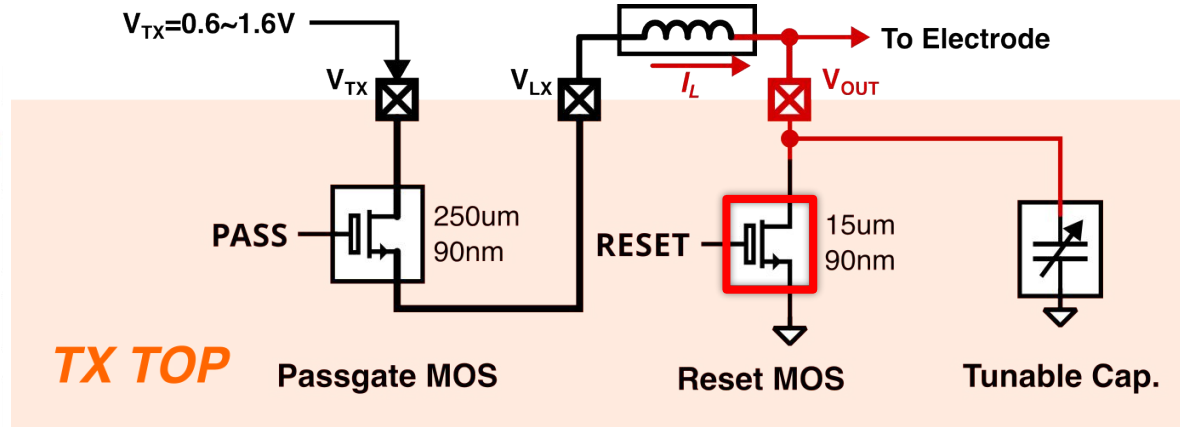
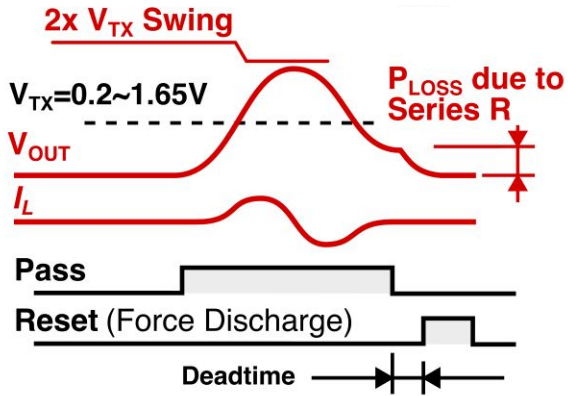


Resonant Pulse Transmitter: Simplified Diagram



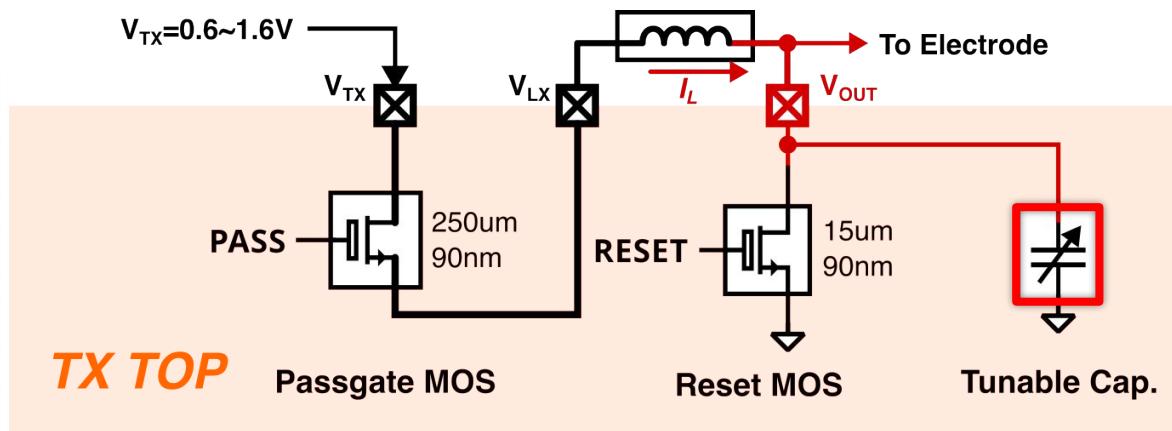
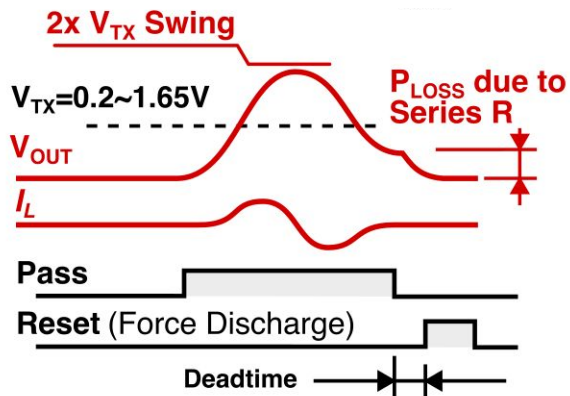
Core switch implemented with a single LVT NMOS 250 μm width to support up to 5mA peak current.

Resonant Pulse Transmitter: Simplified Diagram



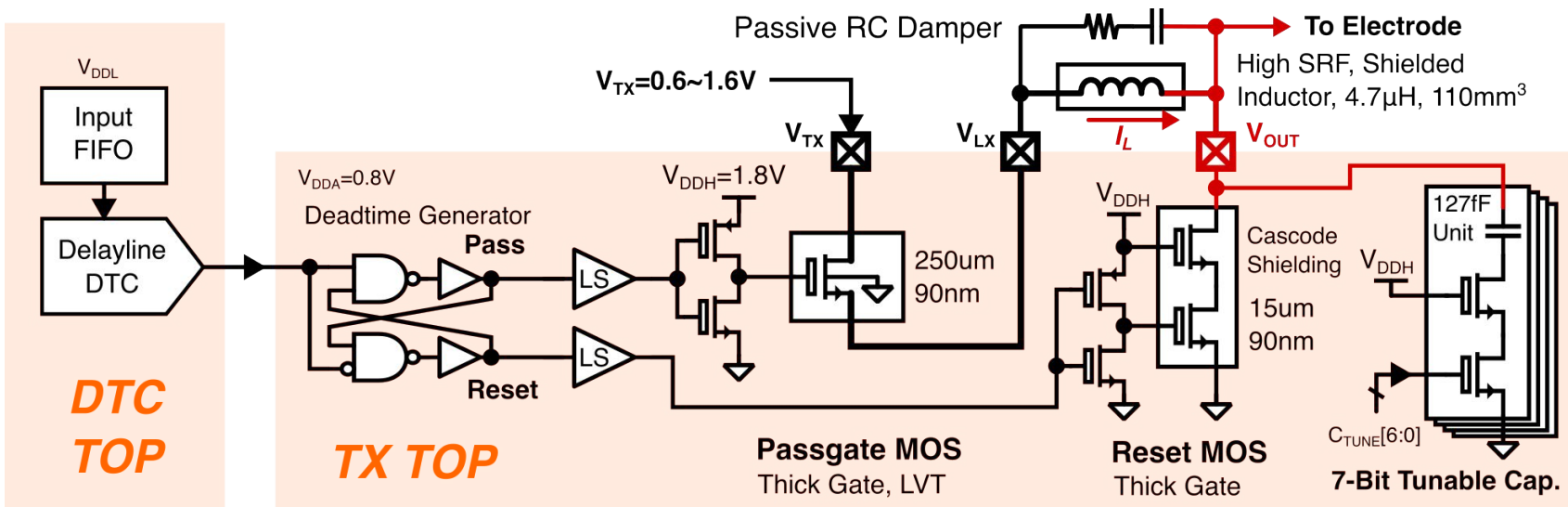
Discharge switch discharges the residue voltage, forces the output to 0V when idle.

Resonant Pulse Transmitter: Simplified Diagram



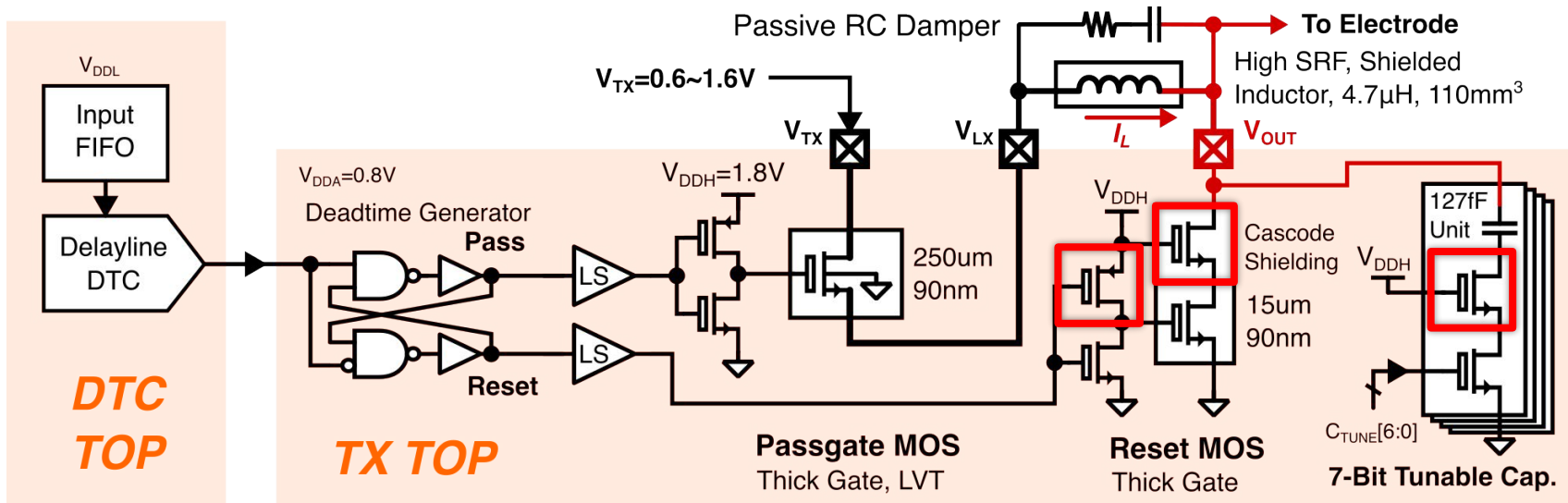
16pF tunable capacitor for resonant frequency trimming

Resonant Pulse Transmitter: Complete Diagram



Complete Circuit Diagram with DTC, Dead-time Generator, and Level-Shifters (LS)

Resonant Pulse Transmitter: Complete Diagram

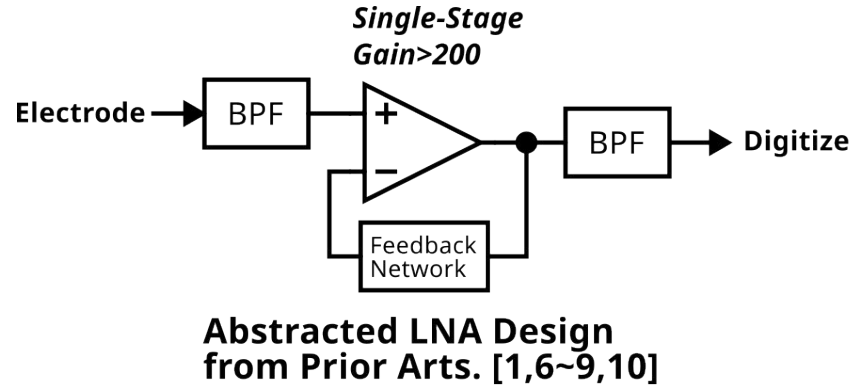
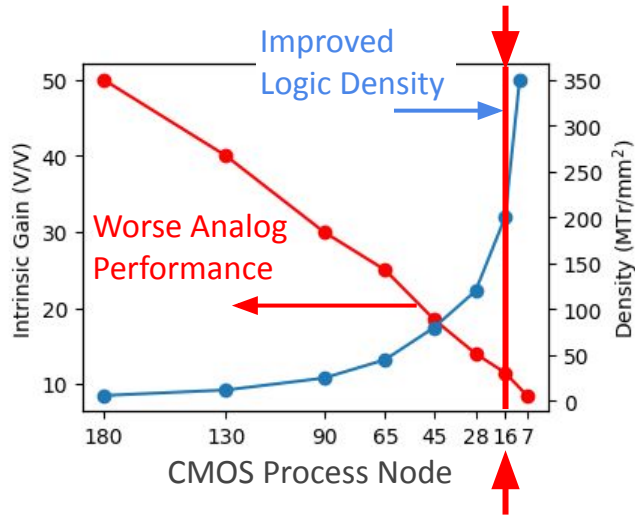


Circuit near V_{OUT} has shielding
to support up to $2 \times V_{DDIO} = 3.6V$ output amplitude

Outline

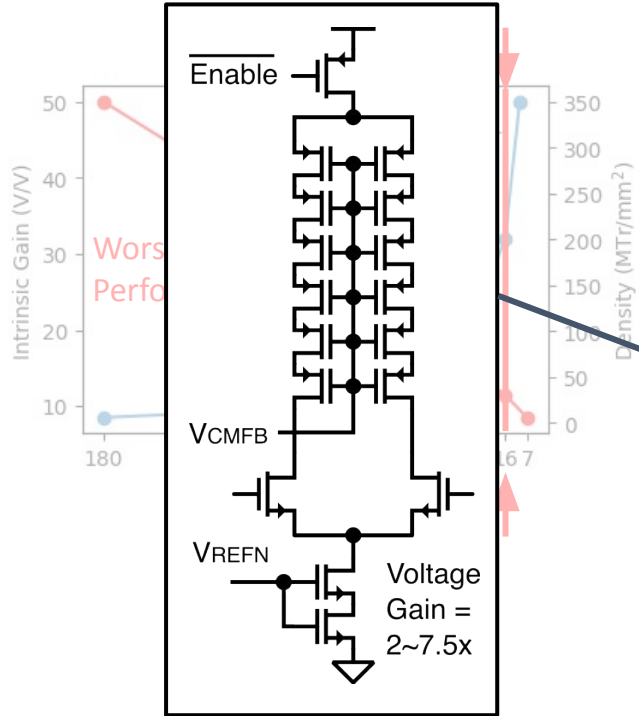
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Low Power Receiver: FinFET LNA Design

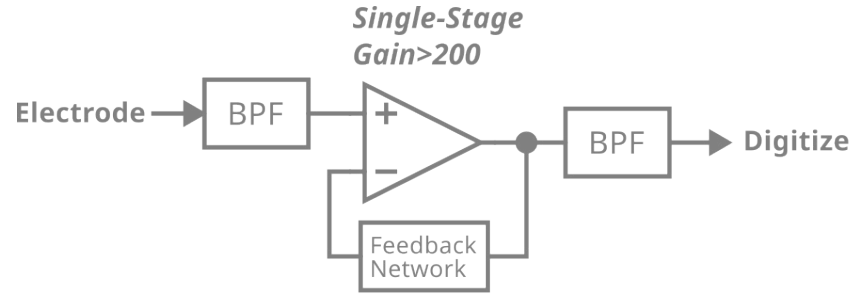


Intrinsic gain in 16nm is only 10V/V !
Hard to achieve high gain with 1 stage.

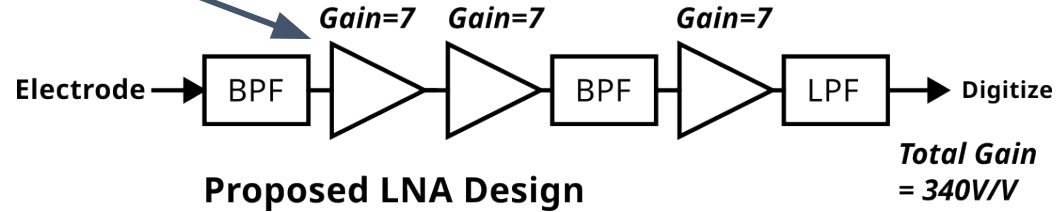
Low Power Receiver: FinFET LNA Design



Basic 5-T OTA Impl. in FinFET

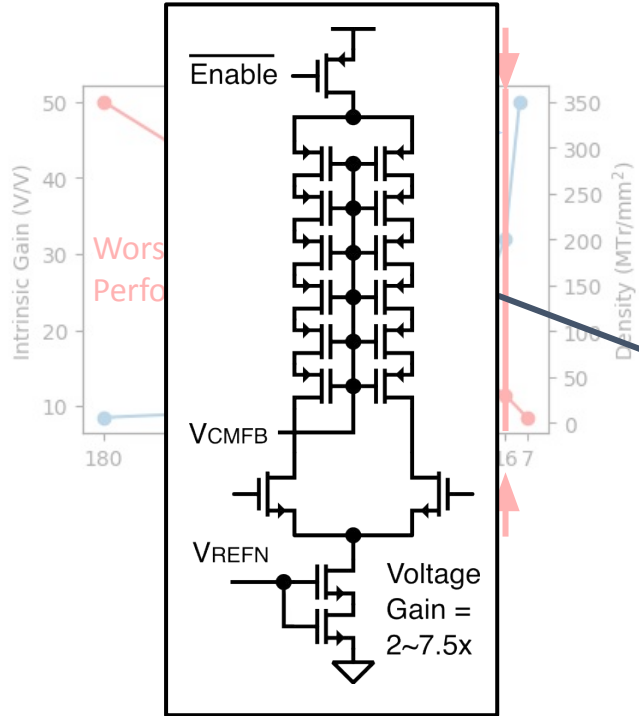


Abstracted LNA Design from Prior Arts. [1,6~9,10]

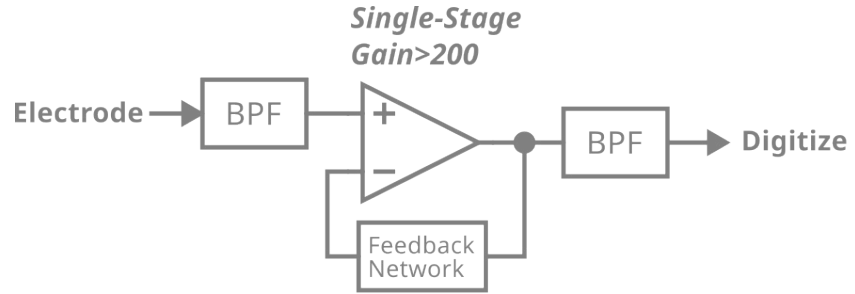


Each stage only need to provide 7V/V gain
 Latency is negligible from a system perspective.

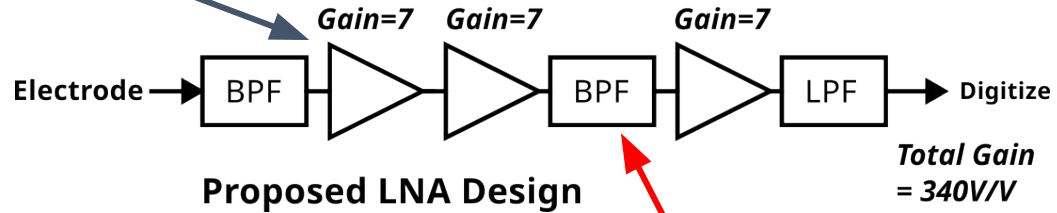
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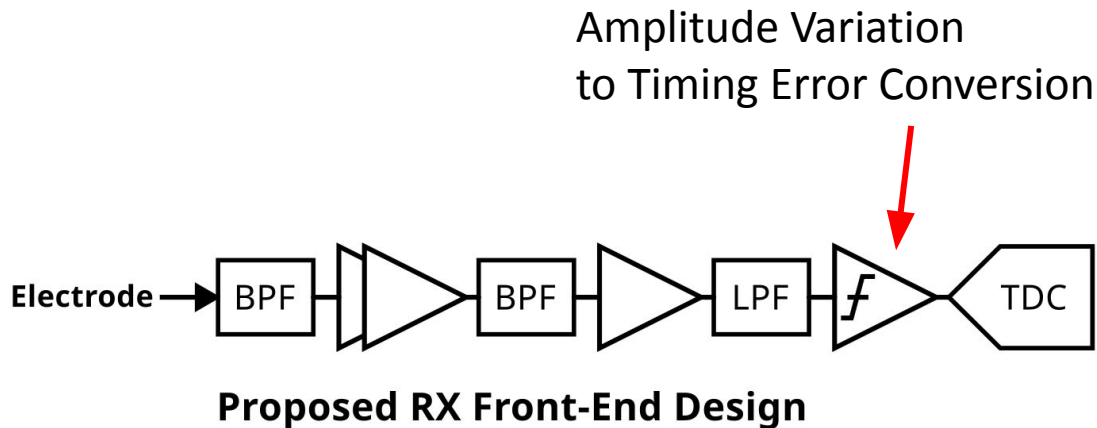
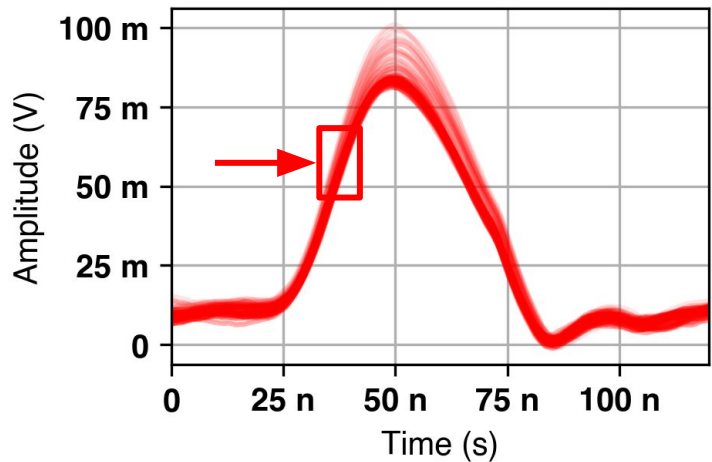
Abstracted LNA Design from Prior Arts. [1,6~9,10]



Proposed LNA Design

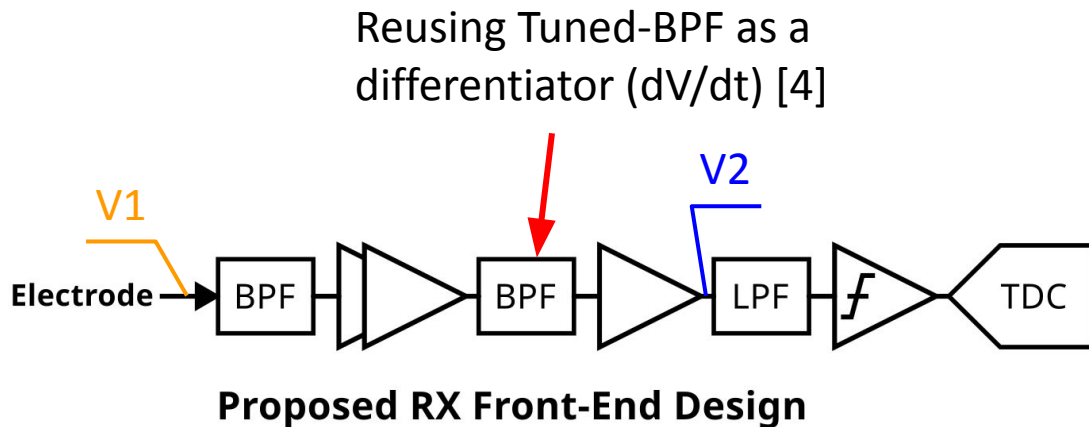
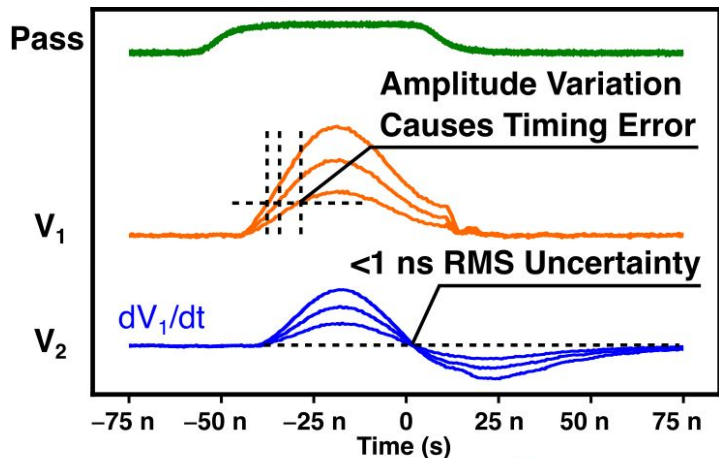
BPF/HPF Removes DC Offset
Offset trimming not required.

Low Power Receiver: Amplitude Variation Rejection



Amplitude variation causes
>4 ns timing error

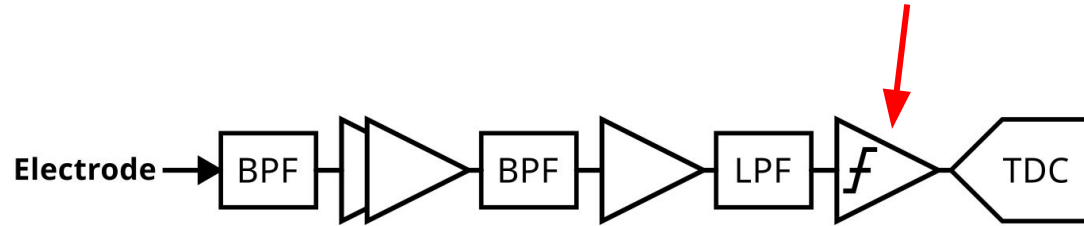
Low Power Receiver: Amplitude Variation Rejection



- ✓ Reduces timing error in practical setups to enable 6b/pulse
- ✗ -3dB SNR Penalty

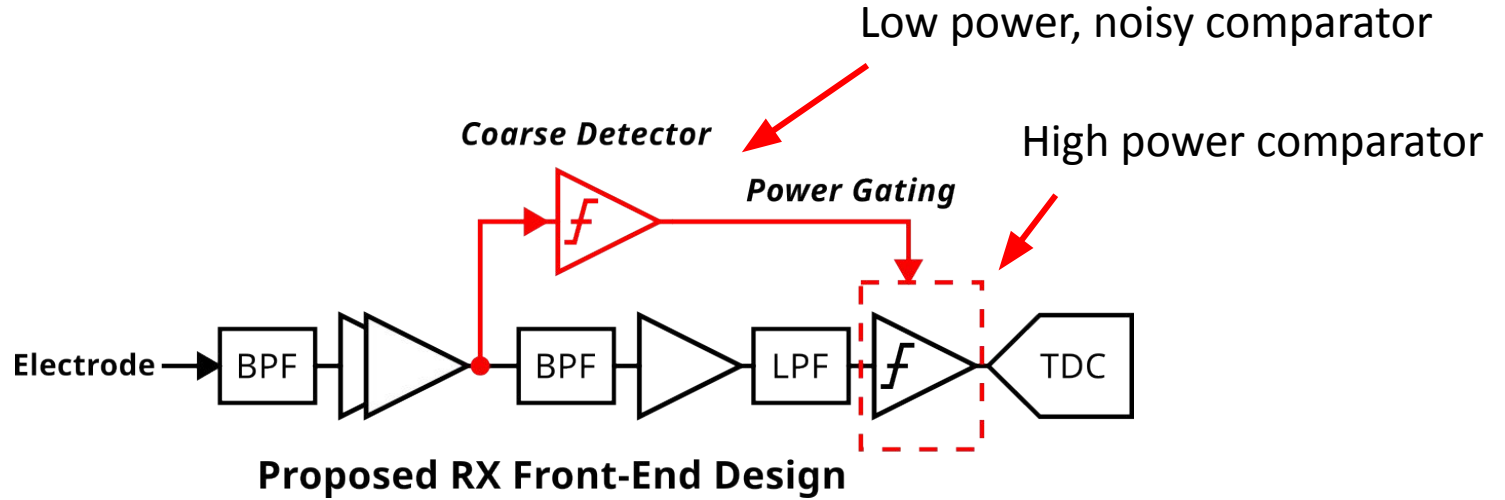
Low Power Receiver: Power Gating

Very high power (20uW)
compared to the other stages



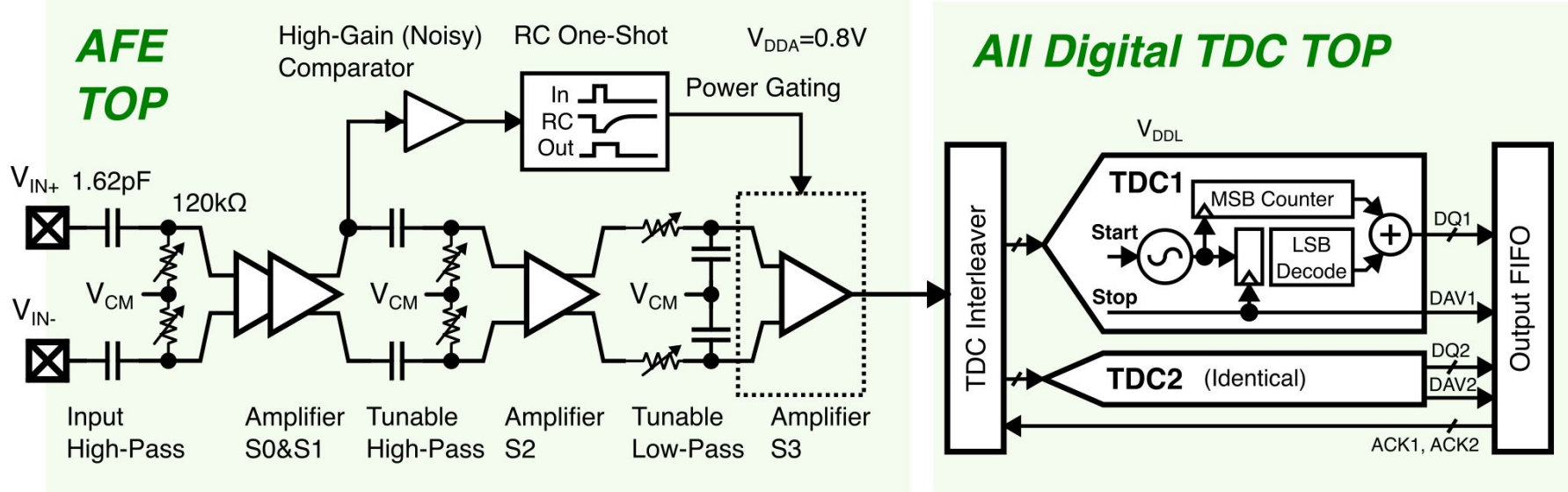
Proposed RX Front-End Design

Low Power Receiver: Power Gating



Proposed Solution: Use delay to our advantage.
Rising edge is non-critical. Use it to power gate the falling edge detector.

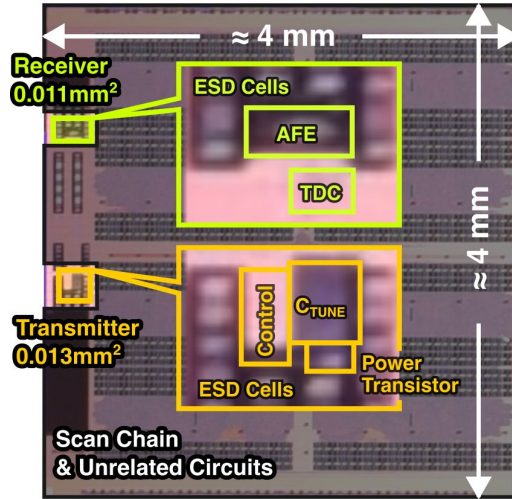
Low Power Receiver: Complete Diagram



Outline

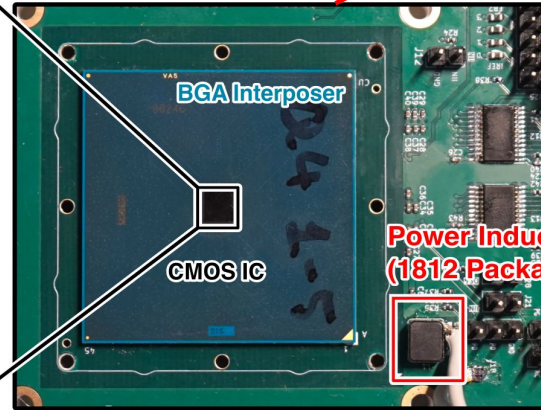
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Chip & Test Board Photograph

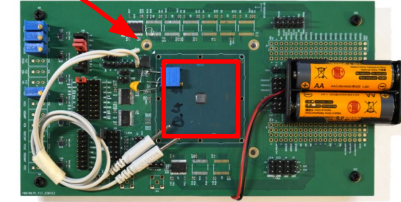


Process Node: 16nm
Package: Flip-Chip BGA

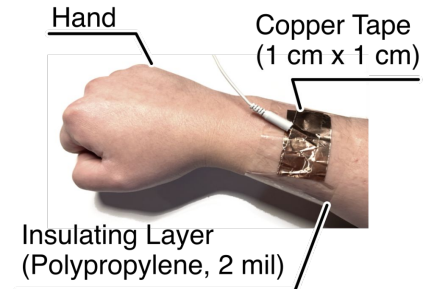
Chip Micrograph & Assembly



PCB Photograph

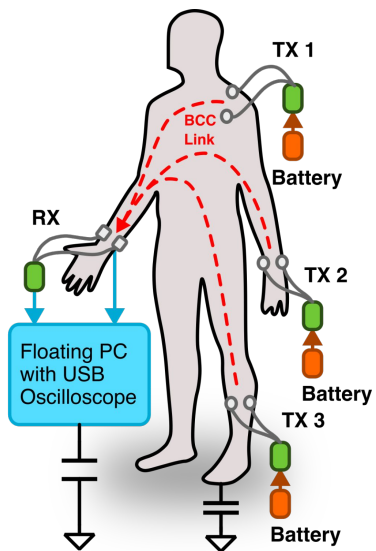


Battery Powered TX Board

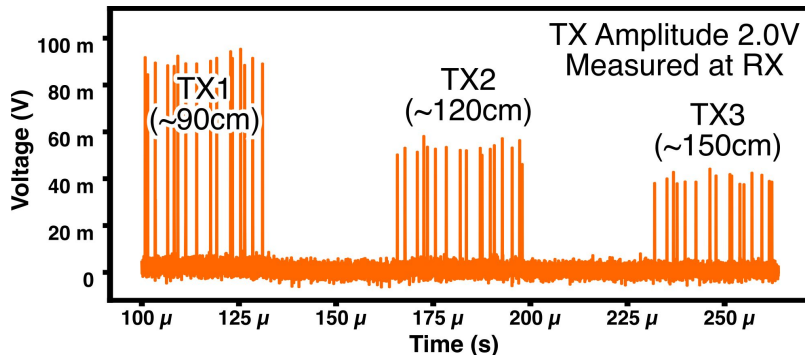


Electrode Structure

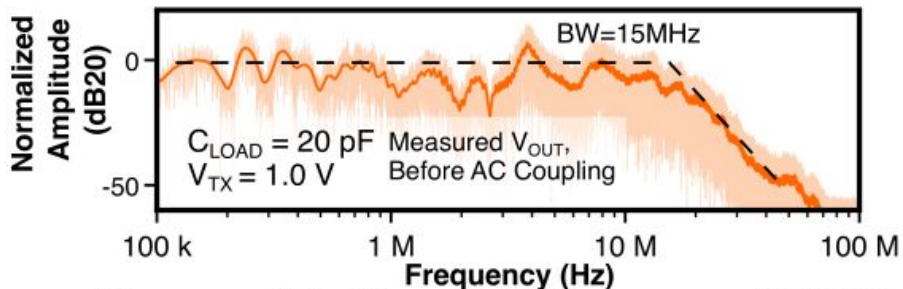
Measurements



Multi-Node TX Setup

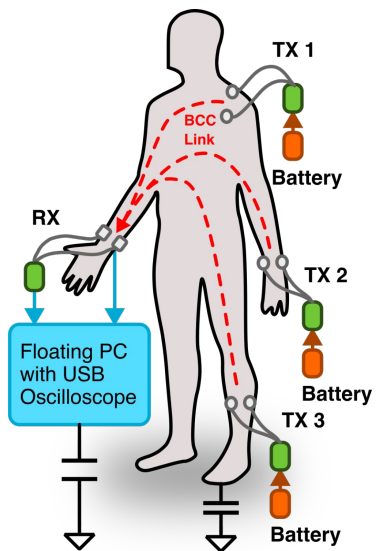


Measured TDMA Waveform of PRBS Data

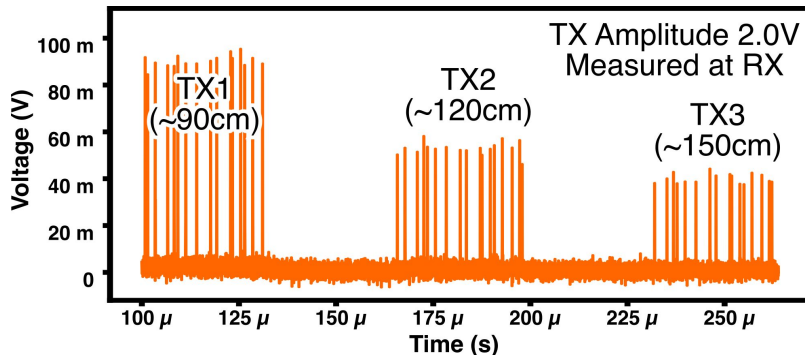


Measured Unfiltered TX Spectrum of PRBS

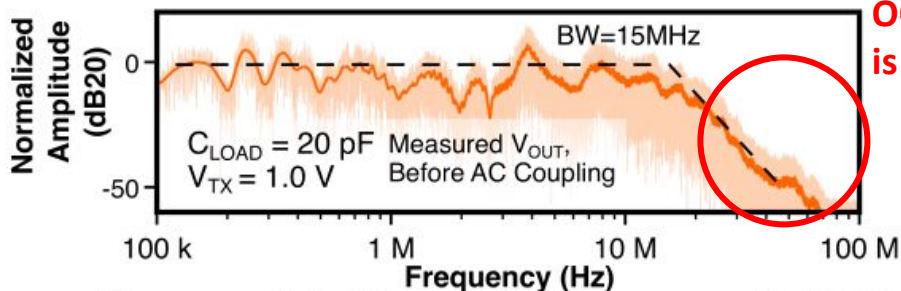
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Multi-Node TX Setup

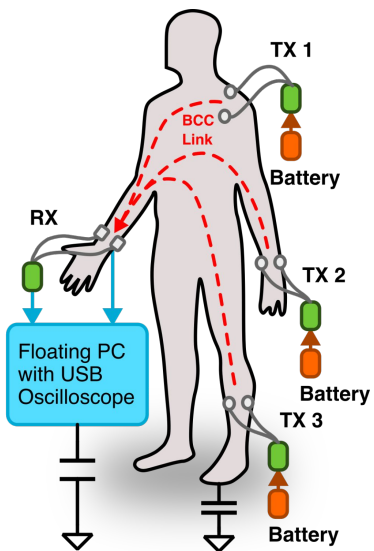


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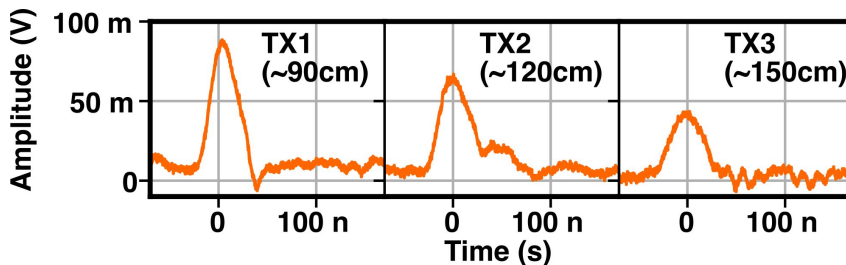


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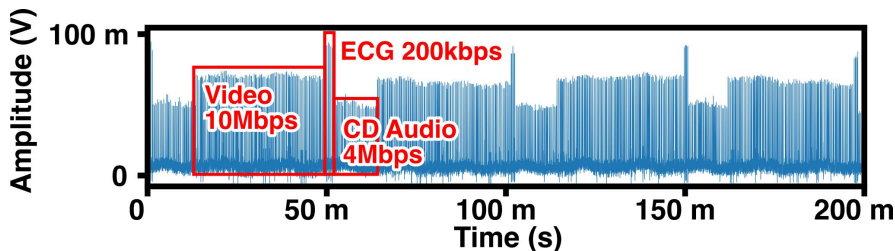
Measurements



Multi-Node TX Setup

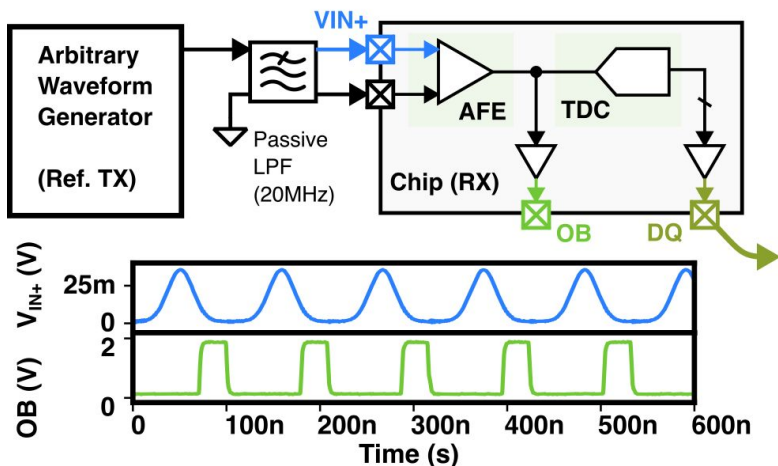


Measured Waveform at RX

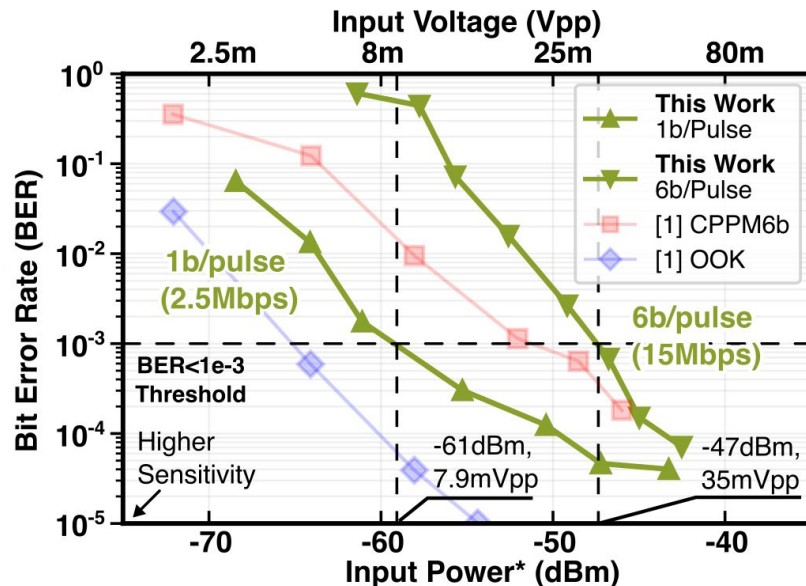


Measured Multi-Node Transaction Demo

Measurements



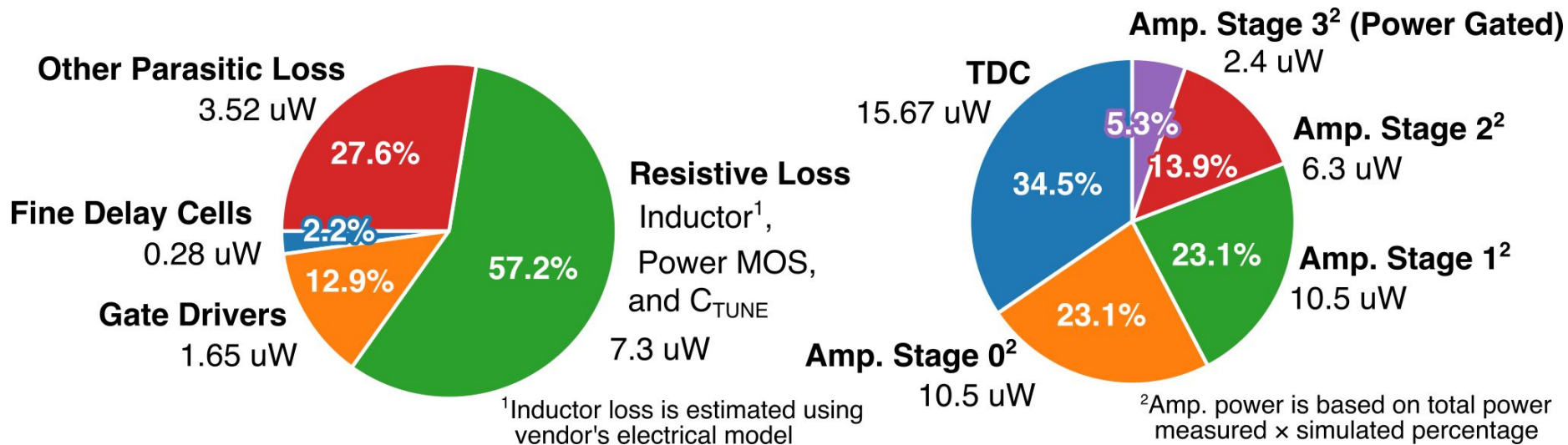
RX Performance Measurement Setup



*Input is matched to 1pF @ 20MHz (~8kΩ Impedance)

Measured RX Sensitivity vs. Input Power

Conclusion: Power Breakdown



Conclusion

Paper	This Work	B.Chatterjee, ISSCC'22 [1]	N. Modak, CICC'21 [6]	S.Maity, JSSC'21 [7]	J. Park, ISSCC'19 [8]	S.Maity, JSSC'19 [9]
Technique	Inductor Resonance	Adiabatic Switching	Inductor Resonance	CMOS Push-Pull	Push-Pull, Magnetically Coupled	CMOS Push-Pull
Modulation	DMPPM6b	CPPM (3~6b)	OOK	OOK	OOK	NRZ
Demod. Method	TDC + DTC	CDR	CDR	CDR	Envelope Detect	CDR
Carrier Freq.	15 MHz	20 MHz	0.5~2 MHz	0.05~1 MHz	40 MHz	-
Data Rate (Mbps)	2.5~15	7.5~15	0.001~0.02	0.001~0.02	5	30
VDD (V)	0.8	0.75	0.5	0.5	0.6	1
Maximum TX Amplitude w/o Stress (V)	3.3	0.7	10	0.5	-	-
Best TX Efficiency (pJ/b)	0.85 (6 b/pulse)	4.2 (NRZ CPPM-6b)	169	277.9	11.2	34.9
Best RX Efficiency (pJ/b)	3.2 (6 b/pulse)	4.0 (NRZ CPPM-6b)	3.6	140	4.8	3.27 (w/o Clock)
TX Power (uW)	12.75	63	1.5	2.47	35.8	2600
RX Power (uW)	48	60	72	1.4	24	98
Area (mm²)	RX: 0.011 TX: 0.013	0.259	0.13	0.3375	0.12	-
Process Node	16nm	65nm	65nm	65nm	65nm	65nm
Sensitivity (dBm)	-47 (6 b/p) -61 (1 b/p)	-65 (OOK) -52 (CPPM)	-60	-64	-56	-64
BER Target	1E-3	1E-3	1E-5	1E-3	1E-3	1E-3

Conclusion

Supports High Data Rate Applications

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BER Target	1E-3	1E-3	1E-5	1E-3	1E-3	1E-3

Conclusion

High TX Amplitude (2x IO transistor breakdown voltage)

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Modulation	DMPPM6b	CPPM (3~6b)	OOK	OOK	OOK	NRZ
Demod. Method	TDC + DTC	CDR	CDR	CDR	Envelope Detect	CDR
Carrier Freq.	15 MHz	20 MHz	0.5~2 MHz	0.05~1 MHz	40 MHz	-
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TX Power (uW)	12.75	63	1.5	2.47	35.8	2600
RX Power (uW)	48	60	72	1.4	24	98
Area (mm ²)	RX: 0.011 TX: 0.013	0.259	0.13	0.3375	0.12	-
Process Node	16nm	65nm	65nm	65nm	65nm	65nm
Sensitivity (dBm)	-47 (6 b/p) -61 (1 b/p)	-65 (OOK) -52 (CPPM)	-60	-64	-56	-64
BER Target	1E-3	1E-3	1E-5	1E-3	1E-3	1E-3

Conclusion

High Power Efficiency: 2.2pJ/b assuming 50% TX, 50% RX

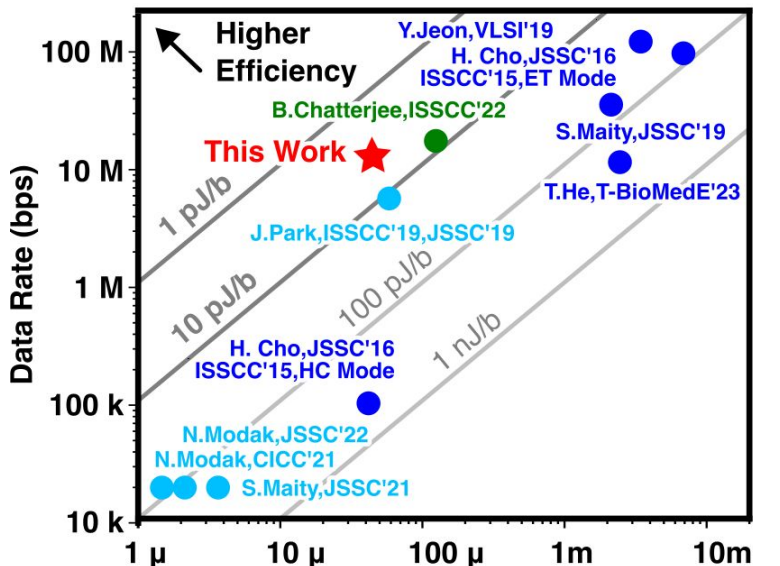
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TX Power (uW)	12.75	63	1.5	2.47	35.8	2600
RX Power (uW)	48	60	72	1.4	24	98
Area (mm2)	RX: 0.011 TX: 0.013	0.259	0.13	0.3375	0.12	-
Process Node	16nm	65nm	65nm	65nm	65nm	65nm
Sensitivity (dBm)	-47 (6 b/p) -61 (1 b/p)	-65 (OOK) -52 (CPPM)	-60	-64	-56	-64
BER Target	1E-3	1E-3	1E-5	1E-3	1E-3	1E-3

Conclusion

Small On-Chip Area

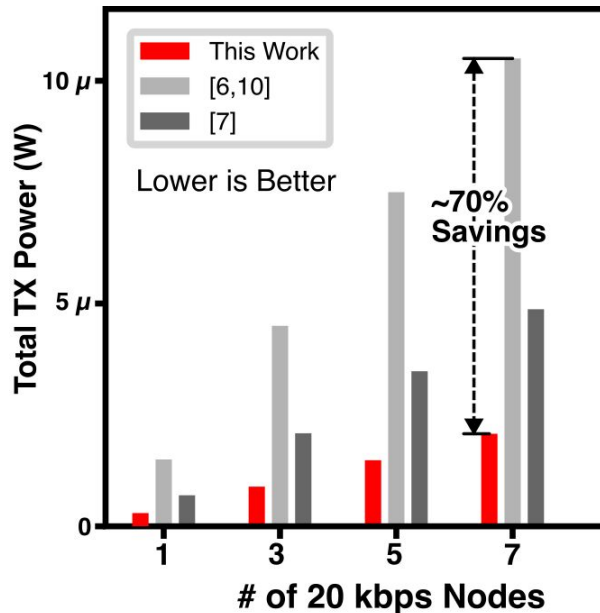
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BER Target	1E-3	1E-3	1E-5	1E-3	1E-3	1E-3

Conclusion



- Pulse Position Modulation (PPM)
- Simple Modulations (OOK/ASK)
- Advanced Modulations (FSK/PAM/etc.)

Energy Efficiency Plot of Recent HBC Transceivers



*Assuming TDMA for this work, FDMA for others; Leakage is included

Power Estimation of Low-Speed Multi-Node Applications

References

1. B. Chatterjee, A. Datta, M. Nath, G. K. K. N. Modak, and S. Sen, "A 65nm 63.3 μ W 15Mbps Transceiver with Switched-Capacitor Adiabatic Signaling and Combinatorial-Pulse-Position Modulation for Body-Worn Video-Sensing AR Nodes," 2022 IEEE International Solid-State Circuits Conference (ISSCC), San Francisco, CA, USA, 2022, pp. 276-278.
2. S. Maity, M. He, M. Nath, D. Das, B. Chatterjee, and S. Sen, "Bio-Physical Modeling, Characterization, and Optimization of Electro-Quasistatic Human Body Communication," in IEEE Transactions on Biomedical Engineering, vol. 66, no. 6, pp. 1791-1802, June 2019.
3. G. Lee, J. Jang, J. -H. Kim and T. W. Kim, "An IR-UWB CMOS Transceiver With Extended Pulse Position Modulation," in IEEE Journal of Solid-State Circuits, vol. 57, no. 8, pp. 2281-2291, Aug. 2022.
4. K. Li, S. Su, Z. Huang, Y. Zhao, and J. Guo, "A Low Walk Error Timing Discrimination ASIC With Rail-to-Rail Dynamic Range and ICMR for Pulsed ToF LiDAR Receiver," in IEEE Transactions on Instrumentation and Measurement, vol. 74, pp. 1-13, 2025, Art no. 2005713.
5. M. Z. Straayer and M. H. Perrott, "A Multi-Path Gated Ring Oscillator TDC With First-Order Noise Shaping," in IEEE Journal of Solid-State Circuits, vol. 44, no. 4, pp. 1089-1098, April 2009, doi: 10.1109/JSSC.2009.2014709.
6. N. Modak et al., "A 65nm Resonant Electro-Quasistatic 5-240 μ W Human Whole-Body Powering and 2.19 μ W Communication SoC with Automatic Maximum Resonant Power Tracking," 2021 IEEE Custom Integrated Circuits Conference (CICC), Austin, TX, USA, 2021, pp. 1-2.
7. S. Maity et al., "Sub- μ WRComm: 415-nW 1-10-kb/s Physically and Mathematically Secure Electro-Quasi-Static HBC Node for Authentication and Medical Applications," in IEEE Journal of Solid-State Circuits, vol. 56, no. 3, pp. 788-802, March 2021.
8. J. Park and P. P. Mercier, "17.6 A Sub-40 μ W 5Mb/s Magnetic Human Body Communication Transceiver Demonstrating Trans-Body Delivery of High-Fidelity Audio to a Wearable In-Ear Headphone," 2019 IEEE International Solid-State Circuits Conference - (ISSCC), San Francisco, CA, USA, 2019.
9. S. Maity, B. Chatterjee, G. Chang, and S. Sen, "BodyWire: A 6.3-pJ/b 30-Mb/s -30-dB SIR-Tolerant Broadband Interference-Robust Human Body Communication Transceiver Using Time Domain Interference Rejection," in IEEE Journal of Solid-State Circuits, vol. 54, no. 10, pp. 2892-2906, Oct. 2019.
10. N. Modak et al., "EQS Res-HBC: A 65-nm Electro-Quasistatic Resonant 5-240 μ W Human Whole-Body Powering and 2.19 μ W Communication SoC With Automatic Maximum Resonant Power Tracking," in IEEE Journal of Solid-State Circuits, vol. 57, no. 3, pp. 831-844, March 2022.
11. H. Cho et al., "A 79 pJ/b 80 Mb/s Full-Duplex Transceiver and a 42.5 μ W 100 kb/s Super-Regenerative Transceiver for Body Channel Communication," in IEEE Journal of Solid-State Circuits, vol. 51, no. 1, pp. 310-317, Jan. 2016.
12. H. Cho, H. Kim, M. Kim, J. Jang, J. Bae, and H.-J. Yoo, "21.1 A 79pJ/b 80Mb/s full-duplex transceiver and a 42.5 μ W 100kb/s super-regenerative transceiver for body channel communication," 2015 IEEE International Solid-State Circuits Conference - (ISSCC) Digest of Technical Papers, San Francisco, CA, USA, 2015, pp. 1-3.
13. T. He et al., "A Re-Configurable Body Channel Transceiver Towards Wearable and Flexible Biomedical Sensor Networks," in IEEE Transactions on Biomedical Circuits and Systems, vol. 17, no. 5, pp. 1022-1034, Oct. 2023.
14. Y. Jeon et al., "A 100Mb/s Galvanically-Coupled Body-Channel-Communication Transceiver with 4.75pJ/b TX and 26.8 pJ/b RX for Bionic Arms," 2019 Symposium on VLSI Circuits, Kyoto, Japan, 2019, pp. C292-C293.

